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LM4930 Boomer™ Audio Power Amplifier Series Audio Subsystem with Stereo Headphone & Mono Speaker Amplifiers

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FEATURES

- 16-bit Resolution 48kHz Stereo DAC
- 16-bit Resolution 8kHz Voice Codec
- I²S Digital Audio Data Serial Interface
- Two-wire Serial Control Interface
- PCM Voice Audio Data Serial Interface
- 25mW/channel Stereo Headphone Amplifier
- 330mW Mono 8 Ω Amplifier (at AV_{DD} = 3.0V)
- 32-step Volume Control for Audio Output Amplifiers
- No Snubber Networks or Bootstrap Capacitors are Required by the Headphone or Hands-free Amplifiers
- Digital Sidetone Generation with Adjustable Attenuation
- Gain Controllable Headphone Amp, Mono BTL Amp, Mic Preamp
- Available in the 36-bump DSBGA and 44-lead WQFN Packages

APPLICATIONS

- Mobile Phones
- Mobile/low Power Audio Appliances
- PDAs

KEY SPECIFICATIONS

- PLS OUT at AV_{DD} = 5.0V, 8Ω 1% THD+N 1W (typ)
- PLS OUT at AV_{DD} = 3.0V, 8Ω 1% THD+N 330mW (typ)
- PH/P OUT at AV_{DD} = 3.0V, 32Ω 0.5% THD+N 25mW (typ)
- Supply Voltage Range
 - DV_{DD}⁽¹⁾ 2.6V to 4.5V
 - AV_{DD}⁽¹⁾ 2.6V to 5.5V
- Total Shutdown Current 2µA (typ)
- PSRR at 217Hz, AV_{DD} = 3V 50dB (typ)
- (1) Best operation is achieved by maintaining 3.0V ≤ AV_{DD} ≤ 5.0 and 3.0V ≤ DV_{DD} ≤ 3.6V. AV_{DD} must be equal to or greater than DV_{DD}, for proper operation.

DESCRIPTION

The LM4930 is an integrated audio subsystem that supports voice and digital audio functions. The LM4930 includes a high quality I²S input stereo DAC, a voice band codec, a stereo headphone amplifier and a high-power mono speaker amplifier. It is primarily designed for demanding applications in mobile phones and other portable devices.

The LM4930 features an l^2S serial interface for full range audio, a 16-bit PCM bi-directional serial interface for the voice band codec and an two-wire interface for control. The full range music path features an SNR of 86dB with a 16-bit 48kHz input. The stereo DAC can also be used while the voice codec is in use. The headphone amplifier delivers 25mW_{RMS} to a 32Ω single-ended stereo load with less than 0.5% distortion (THD+N) when $AV_{DD}=3V$. The mono speaker amplifier delivers up to 330mW into an 8Ω load with less than 1% distortion when $AV_{DD}=3V$.

The LM4930 employs advanced techniques to reduce power consumption, to reduce controller overhead and to eliminate click and pop. Boomer audio power amplifiers were designed specifically to provide high quality output power with a minimal amount of external components. It is, therefore, ideally suited for mobile phone and other low voltage applications where minimal power consumption is a primary requirement.



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Typical Application

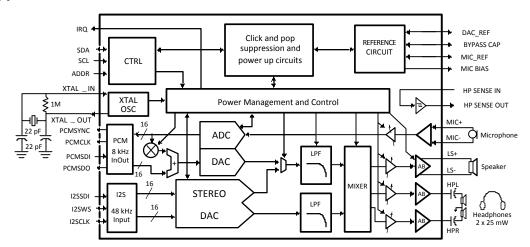
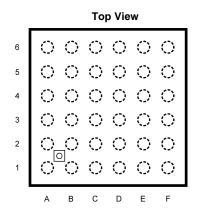


Figure 1. Typical I²S + Voice Codec Application Circuit for Mobile Phones

Connection Diagrams



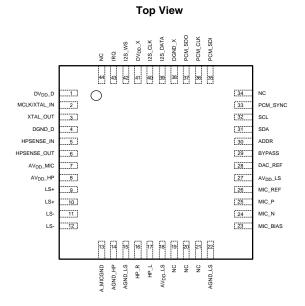


Figure 2. 36 - Bump DSBGA Package See Package Number YZR0036KRA

Figure 3. 44 - Lead WQFN Package See Package Number NJN0044A



Pin Descriptions

Pin No.	Pin Name	Description
A1	MIC_P	Microphone positive differential input
A2	MIC_N	Microphone negative differential input
А3	AVDD_MIC	Analog V _{dd} for microphone preamp
A4	DAC_REF	D/A converter reference voltage
A5	SDA	Two-wire control interface serial data pin
A6	SCL	Two-wire control interface serial clock pin
B1	AGND_MIC	Analog ground for microphone preamp
B2	MIC_BIAS	Microphone bias supply output (2V)
В3	MIC_REF	Internal fixed-reference bypass capacitor decoupling pin
B4	ADDR	Control bus address select pin
B5	PCM_SDI	PCM serial data in
B6	PCM_CLK	PCM Serial clock pin
C1	AVDD_HP	Analog V _{dd} for headphone amplifier
C2	NC	No Connect
C3	BYPASS	Half-supply bypass capacitor decoupling pin
C4	PCM_SYNC	PCM Frame sync pin
C5	I2S_DATA	I ² S serial data input
C6	DGND_D	Digital ground
D1	HP_L	Headphone amplifier connection (Left)
D2	HP_R	Headphone amplifier connection (Right)
D3	HPSENSE_IN	Connection for sense pin of headphone jack
D4	PCM_SDO	PCM serial data out
D5	I2S_CLK	I ² S serial bit clock
D6	DVDD_D	Digital V _{dd}
E1	AGND_HP	Analog ground for headphone amplifier
E2	LS-	Loudspeaker amplifier BTL negative out (-)
E3	HPSENSE_OUT	Logic output pin to indicate headphone connection status. Outputs logic high when HPSENSE_IN is high and outputs logic low when HPSENSE_IN is low. See Figure 50 for suggested application circuit
E4	IRQ	LM4930 mode status indicator pin
E5	I2S_WS	I ² S word select
E6	XTAL_OUT	Negative feedback source for external crystal MCLK
F1	AGND_LS	Analog ground for loudspeaker amplifier
F2	LS+	Loudspeaker amplifier BTL positive out (+)
F3	AVDD_LS	Analog V _{DD} for loudspeaker amplifier
F4	DGND_X	Digital ground
F5	DVDD_X	Digital V _{DD}
F6	MCLK/XTAL_IN	12.288MHz or 24.576MHz Master Clock from crystal (via XTAL OUT) or external source

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System Control Registers

The LM4930 is controlled with a two-wire serial interface. This interface is used to configure the operating mode, digital interfaces, and delta-sigma modulators. The LM4930 is controlled by writing information into a series of write-only registers, each with its own unique 7 bit address. The following registers are programmable:

Table 1. BASIC CONFIGURATION Registers (1)

BASIC CO	NFIGURATIO	ON (XX1000). (Set =	logic 1	, Clear	= logic	: 0)											
ВІТ	15	14	13	12	11	10	9	8	7	6	5	4	3	2	2	1	0	
RESET	0	0	0	0	0	0	0	0	0	0	0	0	0	(0	0	0	
Address	Register		Desc	ription		1		'				<u>'</u>	<u>'</u>					
3:0	MODE		mode	The LM4930 can be placed in one of several modes that dictate the basic operation mode is selected the LM4930 will change operation silently and will re-configure the management profile automatically. The modes are described as follows: (2)													new	
			Mode	•		Speal			Headphone Left Source			lphon ce	e Righ	t (Comn	nent		
			0000		None			None			None	!		F	Powe	rdown	mode	
			0001	01 None			None			None	!		5	Stand	by mod	de		
				0010 Voice			None			None	1			Mono mode	speak	ər		
					None			Voice			Voice)			Headp mode	ohone (call	
		0100		Voice	!		Voice			Voice)			Conference call mode		call		
			0101		Audio	(L+R)		None	None			None			L+R n speak		o mono	
			0110	None			Audio	(Left)		Audio	(Rigl	ht)		Headp audio	ohone s	stereo		
			0111		Audio	(L+R)		Audio	Audio (Left) Voice			Audio (Right) Voice			L+R mixed to mor speaker + stereo headphone audio		ereo	
			1000		Audio	(Left)		Voice							Mixed Mode			
			1001		Voice	+ Audi	o (Left)	Voice	Voice			Voice			Mixed mode			
			1010		Voice	!		Audio	Audio (Left)			Audio (Left)				Mixed Mode		
4	SOFT_RES	SET	Rese	ts the L	M4930,	exclud	ing the	control	registe	'S								
5	PCM_LONG	G	If set	the PC	M interf	ace use	es a lor	g frame	sync.(3)								
6	PCM_COM	IPANDED	If set	the 8 M	ISBs ar	e presu	med to	be com	panded	d data	and the	8 LSE	3s are i	gnor	ed. (3)			
7	PCM_LAW		If set,	the cor	mpande	ed G71	1 data i	s set to	be A-la	w, else	μ-law i	s assi	umed ⁽³⁾)				
8:9	PCM_SYNO	C_MODE	Sets latter	1 (00h), frames	2 (01h) or 4(1	0h) 16	bit fram	es per s	sync. T	he PCN	/LSD	O pin is	tri-s	tated	during	the	
10	PCM_ALW	AYS_ON	This b	oit shou and sy	ld be se	et if and als in al	ther co	dec is u s excep	sing the	e PCM down ^{(;}	bus. W	hen s	et, the	LM4	930 w	ill drive	e the	
11	I2S_M/S		I2S m	naster o	r slave	select.	If set th	en I2S	= maste	er. Clea	ared = s	lave						
12	I2S_RES		I2S re	esolutio	n select	. If set	then 32	bits pe	r frame	. If clea	ared the	n 16 l	oits per	fram	ne			
13	RSVD		RESE	RVED	(4)													
14	RSVD		RESE	RVED	(4)													
15	RSVD		RESE	RVED	(4)													

This register is used to configure the I2S and PCM interfaces as well as the 48kHz DAC module. The 7 bit address for the BASICCONFIG register is XX10000.

⁽X = 0 if ADDR is set to logic 0)

⁽X = 1 if ADDR is set to logic 1)
With the exception of Standby Mode, rapid switching between modes should be avoided. Rapid switching between modes will not ensure that the desired mode will be activated.

It is recommended to alter this bit only while the part is in Powerdown Mode.

Reserved bits should be set to zero when programming the associated register.



Table 2. VOICE/TEST CONFIG Registers (1)

VOICETES	TCONFIG (X	X10001). (Set = lo	gic 1, C	lear =	logic 0))										
BIT	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
RESET	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
Address	Register		Desc	ription													
0	CLASS			configu cer outp				with an	externa	l class	D or line	ear am	nplifier a	nd turi	ns the B1	L	
4:1	SIDESTON	E_ATTEN	Progr	ams the	attenu	ation c	of the c	ligital sid	detone.	Attenu	ation is s	set as	follows:				
			4:1		Sidet Atten	one luation	ı	4:1	4:1			one A	Attenua	ion			
			0000		Mute			1000)		-9dB						
			0001		-30dE	3		1001			-6dB						
			0010		-27dB			1010)		-3dB						
			0011		-24dE	3		1011			0dB						
			0100		-21dE	3		1100)		Mute						
			0101		-18dE	3		1101	l		Mute						
			0110		-15dE	3		1110			Mute						
			0111		-12dE	3		1111			Mute						
5	AUTOSIDE		may r	not be d	esirabler (0010	e. If set 0, 0100	t, the s , 1001	idetone , and 10	is alway	ys mute	ed in mo	des w	hen voi	ce is p	nere side layed on hatever	the	
6	CLOCK_DI	V	If set,	allows	for the	use of	a 24.5	76MHz	crystal.	Defaul	t setting	is for	12.288	ЛHz cr	ystal. ⁽²⁾		
7	ZXD_DISA	BLE	Disab	les the	zero cr zero cr	ossing oss. ⁽³⁾	detect	in the s	stereo D	AC to	ensure in	mmed	iate mod	de cha	nges rath	ner than	
8:9	RSVD		RESE	RVED ⁽	4)												
10:11	CAP_SIZE			accom					citor val	ues to	give cor	rect tu	ırn-off d	elay ar	nd click/p	ор	
			10:11		Delay	/		Вур	ass Cap	acitor	Size						
			00		25ms			0.1µ	F								
			01		50ms	i		0.39	μF								
			10		85ms	i		1µF									
			11		RESE	RVED		RES	ERVED								
12	ZXDS_SLO	W	If set,	If set, this forces the stereo DAC outputs to wait for a zero crossing before powering down													
13	MUTE_LS		If set,	If set, mutes the loudspeaker amplifier in any mode where it is not already muted													
14	MUTE_HP		If set,	mutes	the hea	adphon	e amp	lifier in a	any mod	le whei	e it is no	ot alre	ady mut	ed			
15	MUTE_MIC	:	If set,	mutes	the mic	rophor	e prea	ımp									

This register configures the voiceband codec, sidetone attenuation, and selected control functions. The 7 bit address for the VOICE TESTCONFIG register is XX10001.

⁽X = 0 if ADDR is set to logic 0)
(X = 1 if ADDR is set to logic 1)
It is recommended to alter this bit only while the part is in Powerdown Mode.
To ensure a successful transistion into Powerdown Mode, ZXD_DISABLE must be set whenever there is no audio input signal present.
Reserved bits should be set to zero when programming the associated register.



Table 3. GAIN CONFIG Registers (1)

0.4.11.001.1	FIG (10/400	10) (0 (le 3. G		CONF	IG Re	egiste	ers							
		10). (Set =				-	0	0	7		-	4	2	0	4		
BIT	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
RESET	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	
Address	Register		Desc	ription	l												
4:0	LOUDSPI	KR_GAIN	Prog	Programs the gain of the loudspeaker amplifier. Gain is set as follows:													
			4:0	4:0		speake	er Gain	4:0			Loud	dspeak	er Gai	n			
				0	-34.50	dB		1000	00		-10.5	dB					
			0000	1	-33dE	3		1000)1		-9dB						
			0001	0	-31.50	dB		1001	0		-7.50	IB					
			0001	00011 -30				1001	1		-6dB						
		0010	0	-28.5dB			1010	10100			IB						
			0010	1	-27dB			10101			-3dB						
			0011	0	-25.5dB			1011	0		-1.50	-1.5dB					
			0011	1	-24dE	-24dB		1011	10111			0dB					
			0100	0	-22.50	dB		1100	00		1.5dl	3					
			0100	1	-21dE	3		1100)1		3dB						
			0101	0	-19.50	dB		1101	0		4.5dl	3					
			0101	1	-18dE	3		11011			6dB						
			0110	0	-16.50	dB		1110	11100			7.5dB					
			0110	1	-15dE	3		1110)1		9dB						
			0111	0	-13.50	dB		1111	0		10.5	dΒ					
			0111	1	-12dE	3		1111	1		12dE	3					
9:5	HP_GAIN		Progi	rams th	e gain o	f the he	eadphor	ne amp	olifier. (Gain is	set as fo	llows:					
			9:5		Head	phone	Gain	9:5			Head	phone	e Gain				
			0000	0	-46dE	3		1000	00		-22.5	idB					
			0000	1	-45dE	3		1000)1		-21d	В					
1			0001	0	-43.50	dB		1001	0		-19.5	idB					
			0001	1	-42db			1001	1		-18d	В					
	1							+									

	01101	-27dB	11101	-3dB	
		-28.5dB			
	01100		11100	-4.5dB	
	01011	-30dB	11011	-6dB	
	01010	-31.5dB	11010	-7.5dB	
	01001	-33dB	11001	-9dB	
	01000	-34.5dB	11000	-10.5dB	
	00111	-36dB	10111	-12dB	
	00110	-37.5dB	10110	-13.5dB	
	00101	-39dB	10101	-15dB	

10100

00100

-40.5dB

(X = 1 if ADDR is set to logic 1)

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-16.5dB

⁽¹⁾ This register is used to control the gain of the headphone amplifier, the loudspeaker amplifier, and the microphone preamplifier. The 7 bit address for the GAINCONFIG register is XX10010.

(X = 0 if ADDR is set to logic 0)



Table 3. GAIN CONFIG Registers⁽¹⁾ (continued)

		13:10	Mic Preamp Gain
		0000	17dB
		0001	19dB
		0010	21dB
		0011	23dB
		0100	25dB
		0101	27dB
		0110	29dB
		0111	31dB
		1000	33dB
		1001	35dB
		1010	37dB
		1011	39dB
		1100	41dB
		1101	43dB
		1110	45dB
		1111	47dB
15:14	RSVD	RESERVED ⁽²⁾	

⁽²⁾ Reserved bits should be set to zero when programming the associated register.

Timing Diagrams

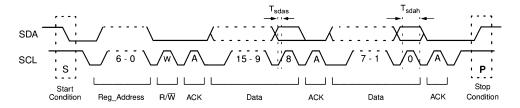


Figure 4. Two-Wire Control Interface Timing Diagram

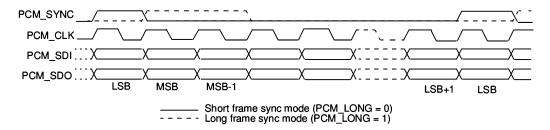


Figure 5. PCM Receive Timing Diagram

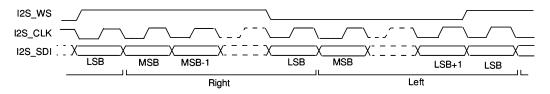


Figure 6. I²S Transmit Timing Diagram

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings (1)(2)(3)

	90	
Analog Supply Voltage		6.0V
Digital Storage Supply Volta	ge	6.0V
Storage temperature		-65°C to +150°C
Power Dissipation (4)		Internally Limited
ESD Susceptibility	Human Body Model ⁽⁵⁾	2000V
	Machine Model ⁽⁶⁾	200V
Junction temperature		150°C
Thermal Resistance	θ _{JA} - YZR0036KRA	105°C/W
	θ _{JA} - NJN0044A ⁽⁷⁾	27°C/W

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which specifies specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication
- All voltages are measured with respect to the relevant GND pin unless otherwise specified.
- If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and
- The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature, T_A . The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$ or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4930, see power derating currents for more information.
- Human body model: 100pF discharged through a $1.5k\Omega$ resistor.
- Machine model: 220pF 240pF discharged through all pins. The given θ_A is for an LM4930 packaged in an NJN0044A with the Exposed-DAP soldered to an exposed 2in^2 area of 1oz printed circuit board copper with 16 thermal vias as described in AN-1187.

Operating Ratings⁽¹⁾

Temperature Range	$T_{MIN} \le T_A \le T_{MAX}$	-30°C ≤ T _A ≤ +85°C
Supply Voltage	DV _{DD} ⁽²⁾	2.6V - 4.5V
	AV _{DD} ⁽²⁾	2.6V - 5.5V

- The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX} , θ_{JA} , and the ambient temperature, T_A . The maximum allowable power dissipation is $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$ or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4930, see power derating currents for more information.
- Best operation is achieved by maintaining 3.0V ≤ AV_{DD} ≤ 5.0 and 3.0V ≤ DV_{DD} ≤ 3.6V. AV_{DD} must be equal to or greater than DV_{DD}. for proper operation.



Electrical Characteristics DV_{DD} = 3.3V, AV_{DD} = 5V, R_{LHP} = 32 Ω , R_{LHF} = 8 $\Omega^{(1)(2)(3)}$

The following specifications apply for the circuit shown in Figure 1, unless otherwise specified. Limits apply for $T_A=25$ °C.

	Devementer	Took Conditions		14930	Units
	Parameter	Test Conditions	Typ ⁽⁴⁾	Limits ⁽⁵⁾⁽⁶⁾	(Limits)
		f _{MCLK} = 12.288MHz			
		Output Mode = "0010" Output Mode = "0011" Output Mode = "0100"	2		
OI _{DD}	Digital Power Supply Current	Output Mode = "0101" Output Mode = "0110" Output Mode = "0111"	4.4		
		Output Mode = "1000" Output Mode = "1001" Output Mode = "1010"	4.9	8	mA (max)
		f _{MCLK} = 12.288MHz; No Load			
		Output Mode = "0010"	7.0		
Analan Daw		Output Mode = "0011"	6.3		
		Output Mode = "0100"	8.0		
AI _{DD}	Analog Power Supply Quiescent Current	Output Mode = "0101"	8.2		
,	Current	Output Mode = "0110"	7.4		
		Output Mode = "0111"	8.7		
		Output Mode = "1000" Output Mode = "1001" Output Mode = "1010"	9.5	14	mA (max)
DI _{SD}	Digital Powerdown Current	f _{MCLK} = 12.288MHz Output Mode = "0000" Powerdown Mode	1	7	μA (max)
AI _{SD}	Analog Powerdown Current	f _{MCLK} = 12.288MHz Output Mode = "0000" Powerdown Mode	1	2	μA (max)
DI _{ST}	Digital Standby Current	f _{MCLK} = 12.288MHz Output Mode = "0001" Standby Mode	1.4	2	mA (max)
Al _{ST}	Analog Standby Current	f _{MCLK} = 12.288MHz Output Mode = "0001" Standby Mode	230	1000	μA (max)
V _{FS_LS}	Full-Scale Output Voltage (Mono speaker amplifier)	CLASS = 0; 0dB gain setting; 8Ω BTL load ⁽⁷⁾	2.5		V _{P-P}
V _{FS_HP}	Full-Scale Output Voltage (Headphone amplifier)	0dB gain setting; 32Ω Stereo Load ⁽⁷⁾	2.5		V _{P-P}
V _{MIC_BIAS}	Mic Bias Voltage		2.0		V
THD+N	Headphone Amplifier Total Harmonic Motion Distortion + Noise	f_{IN} = 1 kHz, P_{OUT} = 7.5mW; 32 Ω Stereo Load	0.07		%
OHP	Headphone Amplifier Output Power	THD+N = 0.5%, f _{OUT} = 1kHz	27	20	mW (min)
Pols	Mono Speaker Amplifier Output Power	THD+N = 1%, f _{OUT} = 1kHz	1		W
PSRR	Power Supply Rejection Ratio	$\begin{array}{l} C_{BYPASS} = 1.0 \mu F \\ C_{DAC_REF} = 1.0 \mu F \\ V_{RIPPLE} = 200 m V_{P.P} @ 217 Hz, MIC_P, \\ MIC_N \ terminated \ with \ 10 \Omega \ to \ ground \end{array}$	55	45	dB (min)

⁽¹⁾ Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which specifies specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication

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 ⁽²⁾ All voltages are measured with respect to the relevant GND pin unless otherwise specified.
 (3) Best operation is achieved by maintaining 3.0V ≤ AV_{DD} ≤ 5.0 and 3.0V ≤ DV_{DD} ≤ 3.6V. AV_{DD} must be equal to or greater than DV_{DD}. for

Typicals are measured at 25°C and represent the parametric norm.

Limits are specified to Texas Instrument's AOQL (Average Outgoing Quality Level).

Datasheet min/max specification limits are ensured by design, test, or statistical analysis.

This value represents the 0dB output level of the given amplifier for the given analog supply voltage. Gain values given in the GAINCONFIG register are relative to these full-scale values for each output amplifier.



Electrical Characteristics DV_{DD} = 3.3V, AV_{DD} = 5V, R_{LHP} = 32Ω , R_{LHF} = $8\Omega^{(1)(2)(3)}$ (continued)

The following specifications apply for the circuit shown in Figure 1, unless otherwise specified. Limits apply for $T_A = 25$ °C.

	Davameter	Toot Complisions	LM	4930	Units
	Parameter	Test Conditions	Typ ⁽⁴⁾	Limits ⁽⁵⁾⁽⁶⁾	(Limits)
SNR (Voice)	Signal-to-Noise Ratio (Voice DAC Path)	Signal = Vo at f = 1kHz @1% THD+N, 32Ω Stereo Load; Noise = digital zero, A- weighted, 0dB gain setting	72		dB
SNR (Music)	Signal-to-Noise Ratio (Music Audio Path)	Signal = Vo at f = 1kHz @1% THD+N, 32Ω Stereo Load; Noise = digital zero, A- weighted; 0dB gain setting	86		dB
DR (Voice)	Dynamic Range (Voice DAC Path)	Signal = Vo at f = 1kHz @1% THD+N, 32Ω Stereo Load; Noise for -60dBFS digital input; A-weighted; 0dB gain setting	72		dB
DR (Music)	Dynamic Range (Music Audio Path)	Signal = Vo at f=1kHz @1% THD+N, 32Ω Stereo Load; Noise for -60dBFS digital input; A-weighted, 0dB gain setting	86		dB
SNR _{ADC}	Signal-to-Noise Ratio (Voice ADC Path)	Reference signal = 0dBFS MIC_P, MIC_N terminated with 10Ω to ground; A-weighted; 47dB MIC preamp gain setting	75		dB
DR _{ADC}	Dynamic Range (Voice ADC Path)	Reference signal = 0dBFS Noise for -60dBFS digital input; A-weighted; 47dB MIC preamp gain setting	75		dB
X _{TALK}	Stereo Channel-to-Channel Crosstalk	$f_S = 48$ kHz, $f_{IN} = 1$ kHz sinewave at -3dB _{FS}	75		dB
V _{MIC-IN}	Maximum Differential MIC Input Voltage	17dB MIC Preamp gain setting	570		mV _{P-P}
R _{VDAC}	Voice DAC Ripple	300Hz - 3.3kHz through head-phone output.	+/-0.15	+/-0.2	dB (max
R _{VADC}	Voice ADC Ripple	300Hz - 3.3kHz through head-phone output.	+/-0.25	+/-0.3	dB (max
PB _{VDAC}	Voice DAC Passband	-3dB Point	3.46		kHz
SBA _{VDAC}	Voice DAC Stopband Attenuation	Above 4kHz	72		dB
UPB _{VADC}	Voice ADC Upper Passband Cutoff Frequency.	Upper -3dB Point	3.47		kHz
LPB _{VADC}	Voice ADC Lower Passband Cutoff Frequency.	Lower -3dB Point	0.230		kHz
SBA _{VADC}	Voice ADC Stopband Attenuation	Above 4kHz	65		dB
SBA _{NOTC}	Voice ADC Notch Attenuation	Centered on 55Hz, figure gives worst case attenuation for 50Hz & 60Hz.	58		dB
R _{DAC}	Audio DAC Ripple	20Hz - 20kHz through head-phone output.	+/-0.1	+/-0.2	dB (max)
PB _{DAC}	Audio DAC Passband Width	-3dB point	22.7		kHz
SBA _{DAC}	Audio DAC Stopband Attenuation	Above 24kHz	76		dB
DR _{DAC}	Audio DAC Dynamic Range Digital Filter Section	Signal = VO at f = 1kHz @ 1% THD+N; f = 1kHz; Noise for -60dBFS digital input; 0dB gain; A-weighted	97		dB
SNR _{DAC}	Audio DAC SNR Digital Filter Section	Signal = VO at f = 1kHz @ 1% THD+N; f = 1kHz; Noise for -60dBFS digital input; 0dB gain; A-weighted	97		dB
ΔA _{CH-CH}	Stereo Channel-to-Channel Gain Mismatch		0.3		dB
V _{IL}	Digital Input: Logic Low Voltage Level		0.4		V
V _{IH}	Digital Input: Logic High Voltage Level		1.4		V
	Volume Control Range (Headphone amplifiers)	Maximum Attenuation Minimum Attenuation	-46.5 0		dB dB
	Volume Control Range (Mono speaker amplifier)	Minimum Gain Maximum Gain	-34.5 12		dB dB

Product Folder Links: LM4930

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Electrical Characteristics DV_{DD} = 3.3V, AV_{DD} = 5V, R_{LHP} = 32Ω , R_{LHF} = $8\Omega^{(1)(2)(3)}$ (continued)

The following specifications apply for the circuit shown in Figure 1, unless otherwise specified. Limits apply for $T_A = 25$ °C.

		Test Conditions			Units	
	Parameter	Test Conditions	Typ ⁽⁴⁾	Limits ⁽⁵⁾⁽⁶⁾	(Limits)	
	Volume Control Step Size (Output amplifiers)		1.5		dB	
	Volume Control Range (Microphone Preamp)	Minimum Gain Maximum Gain	17 47		dB dB	
	Volume Control Step Size (Microphone Preamp)		2		dB	
	Side Tone Attenuation Range	Maximum Attenuation Minimum Attenuation	-30 0		dB dB	
	Side Tone Attenuation Step Size		3		dB	
MCLK	MCLK frequency	CLOCK_DIV = 0 CLOCK_DIV = 1	12.288 24.576		MHz MHz	
	MCLK Duty Cycle		50	40 60	% (min) % (max)	
CONV	Sampling Clock Frequency ⁽⁸⁾		48		kHz	
CLKSCL	SCL_CLK Frequency		400		kHz	
RISESCL	SCL_CLK, SCL_DATA Rise Time		300		ns	
FALLSCL	SCL_CLK, SDA_DATA Fall Time		300		ns	
SDAH	SDA_DATA Hold Time		500		ns	
SDAS	SDA_DATA Setup Time		500		ns	
CLKPCM	PCM_CLK Frequency	PCM_SYNC_MODE = 00 PCM_SYNC_MODE = 01 PCM_SYNC_MODE = 10	128 256 512		kHz	
	PCM_CLK Duty Cycle		50	40 60	% (min) % (max)	
CLKI2S	I2S_CLK Frequency	I2S_RES = 0 I2S_RES = 1	1.536 3.072		MHz	
	I2S_CLK Duty Cycle		50	40 60	% (min) % (max)	

⁽⁸⁾ The sampling clock frequency is equal to the master clock frequency divided by 256. ($f_{conv} = f_{MCLK}/256$)

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Electrical Characteristics DV $_{DD}$ = 3V, AV $_{DD}$ = 3V, R $_{LHP}$ = 32 $\Omega,$ R $_{LHF}$ = 8 $\Omega^{(1)(2)(3)}$

The following specifications apply for the circuit shown in Figure 1, unless otherwise specified. Limits apply for $T_A=25^{\circ}$ C.

	Danier and an	Total Complitions	LN	Units	
	Parameter	Test Conditions	Typ ⁽⁴⁾	Limits ⁽⁵⁾⁽⁶⁾	(Limits)
		f _{MCLK} = 12.288MHz			
		Output Mode = "0010" Output Mode = "0011" Output Mode = "0100"	1.6		
DI _{DD}	Digital Power Supply Current	Output Mode = "0101" Output Mode = "0110" Output Mode = "0111"	3.8		
		Output Mode = "1000" Output Mode = "1001" Output Mode = "1010"	4.2	7	mA (max)
		f _{MCLK} = 12.288MHz; No Load			
		Output Mode = "0010"	5.8		
		Output Mode = "0011"	5.1		
		Output Mode = "0100"	6.5		
AI _{DD}	Analog Power Supply Quiescent Current	Output Mode = "0101"	6.4		
55	Current	Output Mode = "0110"	5.8		
		Output Mode = "0111"	7.0		
		Output Mode = "1000" Output Mode = "1001" Output Mode = "1010"	7.5	12	mA (max)
DI _{SD}	Digital Powerdown Current	f _{MCLK} = 12.288MHz Output Mode = "0000" Powerdown Mode	1	7	μA (max)
AI _{SD}	Analog Powerdown Current	f _{MCLK} = 12.288MHz Output Mode = "0000" Powerdown Mode	0.6	1.5	μA (max)
DI _{ST}	Digital Standby Current	f _{MCLK} = 12.288MHz Output Mode = "0001" Standby Mode	1.1	1.7	mA (max)
Al _{ST}	Analog Standby Current	f _{MCLK} = 12.288MHz Output Mode = "0001" Standby Mode	100	300	μA (max)
V _{FS_LS}	Full-Scale Output Voltage (Mono speaker amplifier)	CLASS = 0; 0dB gain setting; 8Ω BTL load ⁽⁷⁾	2.5		V_{P-P}
V _{FS_HP}	Full-Scale Output Voltage (Headphone amplifier)	0dB gain setting; 32Ω Stereo Load ⁽⁷⁾	2.5		V_{P-P}
V _{MIC_BIAS}	Mic Bias Voltage		2		V
THD+N	Headphone Amplifier Total Harmonic Distortion + Noise	$f_{IN} = 1kHz, P_{OUT} = 7.5mW$	0.07		%
OHP	Headphone Amplifier Output Power	THD+N = 0.5%, f _{OUT} = 1kHz	25	15	mW (min)
P _{OLS}	Mono Speaker Amplifier Output Power	THD+N = 1%, f _{OUT} = 1kHz	330	270	mW (min)
PSRR	Power Supply Rejection Ratio	$C_{BYPASS} = 1.0 \mu F$ $C_{DAC_REF} = 1.0 \mu F$ $V_{RIPPLE} = 200 m V_{P.P} @ 217 Hz$	50	42	dB (min)

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which specifies specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (2) All voltages are measured with respect to the relevant GND pin unless otherwise specified.
- (3) The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{JMAX}, θ_{JA}, and the ambient temperature, T_A. The maximum allowable power dissipation is P_{DMAX} = (T_{JMAX} T_A) / θ_{JA} or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4930, see power derating currents for more information.
- (4) Typicals are measured at 25°C and represent the parametric norm.
- (5) Limits are specified to Texas Instrument's AOQL (Average Outgoing Quality Level).
- (6) Datasheet min/max specification limits are ensured by design, test, or statistical analysis.
- (7) This value represents the 0dB output level of the given amplifier for the given analog supply voltage. Gain values given in the GAINCONFIG register are relative to these full-scale values for each output amplifier.

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Electrical Characteristics DV_{DD} = 3V, AV_{DD} = 3V, R_{LHP} = 32Ω , R_{LHF} = $8\Omega^{(1)(2)(3)}$ (continued)

The following specifications apply for the circuit shown in Figure 1, unless otherwise specified. Limits apply for $T_A = 25$ °C.

	Davameter	Toot Complisions	LM4930		Units
Parameter		Test Conditions	Typ ⁽⁴⁾ Limits ⁽⁵⁾⁽⁶⁾		(Limits)
SNR (Voice)	Signal-to-Noise Ratio (Voice DAC Path)	Signal = Vo at f = 1kHz @1% THD+N, 32Ω Stereo Load; Noise = digital zero, A- weighted; 0dB gain setting	72		dB
SNR (Music)	Signal-to-Noise Ratio (Music Audio Path)	Signal = Vo at f = 1kHz @1% THD+N, 32Ω Stereo Load; Noise = digital zero, A- weighted; 0dB gain setting	86		dB
DR (Voice)	Dynamic Range (Voice DAC Path)	Signal = Vo at f = 1kHz @1% THD+N, 32Ω Stereo Load; Noise for -60dBFS digital input; A-weighted, 0dB gain setting	72		dB
DR (Music)	Dynamic Range (Music Audio Path)	Signal = Vo at f=1kHz @1% THD+N, 32Ω Stereo Load; Noise for -60dBFS digital input; A-weighted, 0dB gain setting	86		dB
SNR _{ADC}	Signal-to-Noise Ratio (Voice ADC Path)	Reference signal = 0dBFS MIC_P, MIC_N terminated with 10Ω to ground; A-weighted; 47dB MIC preamp gain setting	75		dB
DR _{ADC}	Dynamic Range (Voice ADC Path)	Reference signal = 0dBFS Noise for -60dBFS digital input; A-weighted; 47dB MIC preamp gain setting	75		dB
X _{TALK}	Stereo Channel-to-Channel Crosstalk	$f_S = 48$ kHz, $f_{IN} = 1$ kHz sinewave at -3dB _{FS}	73		dB
$V_{\text{MIC-IN}}$	Maximum Differential MIC Input Voltage	17dB MIC Preamp gain setting	570		mV_{P-P}
R _{VDAC}	Voice DAC Ripple	300Hz - 3.3kHz through head-phone output.	+/-0.15	+/-0.2	dB (max
R _{VADC}	Voice ADC Ripple	300Hz - 3.3kHz through head-phone output.	+/-0.25	+/-0.3	dB (max
PB _{VDAC}	Voice DAC Passband	-3dB Point	3.46		kHz
SBA _{VDAC}	Voice DAC Stopband Attenuation	Above 4kHz	72		dB
UPB _{VADC}	Voice ADC Upper Passband Cutoff Frequency.	Upper -3dB Point	3.47		kHz
LPB _{VADC}	Voice ADC Lower Passband Cutoff Frequency.	Lower -3dB Point	0.230		kHz
SBA _{VADC}	Voice ADC Stopband Attenuation	Above 4kHz	65		dB
SBA _{NOTC} H	Voice ADC Notch Attenuation	Centered on 55Hz, figure gives worst case attenuation for 50Hz & 60Hz.	58		dB
R _{DAC}	Audio DAC Ripple	20Hz - 20kHz through head-phone output.	+/-0.1	+/-0.2	dB (max
PB _{DAC}	Audio DAC Passband Width	-3dB point	22.7		kHz
SBA _{DAC}	Audio DAC Stopband Attenuation	Above 24kHz	76		dB
DR _{DAC}	Audio DAC Dynamic Range Digital Filter Section	Signal = VO at f = 1kHz @ 1% THD+N; f = 1kHz; Noise for -60dBFS digital input; 0dB gain; A-weighted	97		dB
SNR _{DAC}	Audio DAC SNR Digital Filter Section	Signal = VO at f = 1kHz @ 1% THD+N; f = 1kHz; Noise for -60dBFS digital input; 0dB gain; A-weighted	97		dB
ΔA _{CH-CH}	Stereo Channel-to-Channel Gain Mismatch		0.3		dB
V _{IL}	Digital Input: Logic Low Voltage Level		0.4		V
V _{IH}	Digital Input: Logic High Voltage Level		1.4		V
	Volume Control Range (Headphone amplifiers)	Maximum Attenuation Minimum Attenuation	-46.5 0		dB dB
	Volume Control Range (Mono speaker amplifier)	Minimum Gain Maximum Gain	-34.5 12		dB dB

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Electrical Characteristics DV_{DD} = 3V, AV_{DD} = 3V, R_{LHP} = 32Ω , R_{LHF} = $8\Omega^{(1)(2)(3)}$ (continued)

The following specifications apply for the circuit shown in Figure 1, unless otherwise specified. Limits apply for $T_A = 25$ °C.

	_	T	LM	LM4930	
Parameter		Test Conditions	Typ ⁽⁴⁾	Limits ⁽⁵⁾⁽⁶⁾	Units (Limits)
	Volume Control Step Size (Output amplifiers)		1.5		dB
	Volume Control Range (Microphone Preamp)	Minimum Gain Maximum Gain	17 47		dB
	Volume Control Step Size (Microphone Preamp)		2		dB
	Side Tone Attenuation Range	Maximum Attenuation Minimum Attenuation	-30 0		dB dB
	Side Tone Attenuation Step Size		3		dB
f _{MCLK}	MCLK frequency	CLOCK_DIV = 0 CLOCK_DIV = 1	12.288 24.576		MHz MHz
	MCLK Duty Cycle		50	40 60	% (min) % (max)
f _{CONV}	Sampling Clock Frequency	See ⁽⁸⁾	48		kHz
f _{CLKSCL}	SCL_CLK Frequency		400		kHz
t _{RISESCL}	SCL_CLK, SCL_DATA Rise Time		300		ns
t _{FALLSCL}	SCL_CLK, SDA_DATA Fall Time		300		ns
t _{SDAH}	SDA_DATA Hold Time		500		ns
t _{SDAS}	SDA_DATA Setup Time		500		ns
f _{CLKPCM}	PCM_CLK Frequency	PCM_SYNC_MODE = 00 PCM_SYNC_MODE = 01 PCM_SYNC_MODE = 10	128 256 512		kHz kHz kHz
	PCM_CLK Duty Cycle		50	40 60	% (min) % (max)
f _{CLKI2S}	I2S_CLK Frequency	I2S_RES = 0 I2S_RES = 1	1.536 3.072		MHz MHz
	I2S_CLK Duty Cycle		50	40 60	% (min) % (max)

⁽⁸⁾ The sampling clock frequency is equal to the master clock frequency divided by 256. ($f_{conv} = f_{MCLK}/256$)



Typical Performance Characteristics (1)

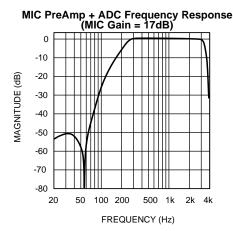
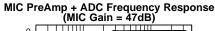


Figure 7.



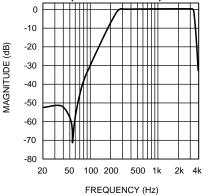


Figure 9.

MIC PreAmp + ADC Frequency Response High Cutoff (MIC Gain = 17dB)

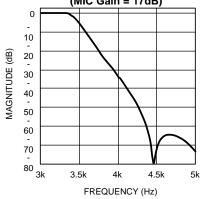


Figure 11.

MIC PreAmp + ADC Frequency Response Zoom (MIC Gain = 17dB)

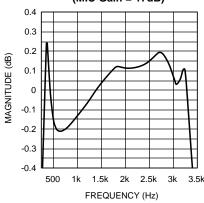


Figure 8.

MIC PreAmp + ADC Frequency Response Zoom (MIC Gain = 47dB)

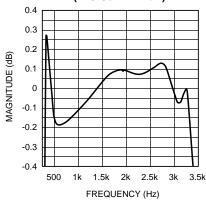


Figure 10.

MIC PreAmp + ADC Frequency Response High Cutoff (MIC Gain = 47dB)

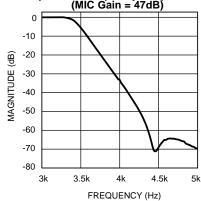
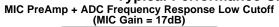
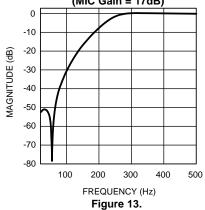


Figure 12.



Typical Performance Characteristics (1) (continued)





ADC THD+N

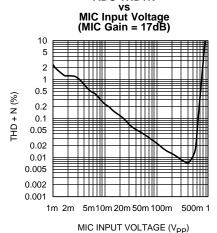


Figure 15.

MIC PreAmp + ADC PSRR vs

Frequency Top Trace = 47dB MIC Gain, Bottom Trace = 17dB MIC Gain

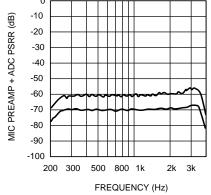
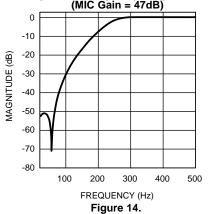


Figure 17.

MIC PreAmp + ADC Frequency Response Low Cutoff (MIC Gain = 47dB)



ADC THD+N

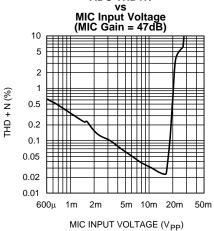


Figure 16.

Headphone Sense In Hysteresis Loop $(AV_{DD} = 3V)$

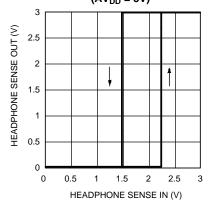


Figure 18.



Typical Performance Characteristics (1) (continued)

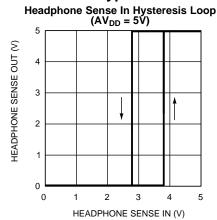
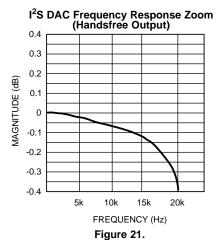


Figure 19.



12S DAC Frequency Response Zoom (Headphone Output)

0.4

0.3

0.2

0.1

-0.1

-0.2

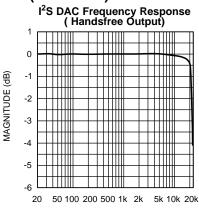
-0.3

-0.4

5k 10k 15k 20k

FREQUENCY (Hz)

Figure 23.



FREQUENCY (Hz) Figure 20.

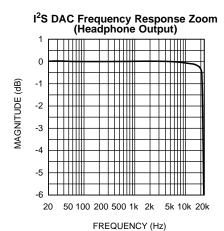
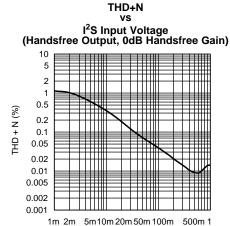


Figure 22.



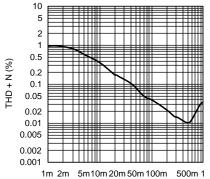
I²S INPUT VOLTAGE (FFS)

Figure 24.



Typical Performance Characteristics (1) (continued) THD+N

I²S Input Voltage (Headphone Output, 0dB Headphone Gain)



I²S INPUT VOLTAGE (FFS)

Figure 25.

MIC Bias Dropout Voltage vs MIC Bias Current

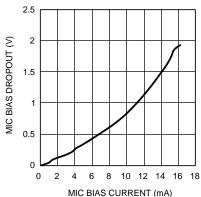


Figure 27.

PCM DAC Frequency Response Zoom (Handsfree Output)

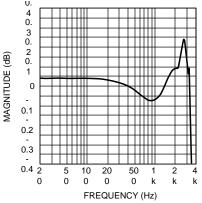
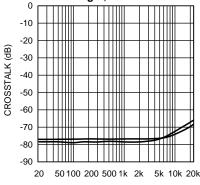


Figure 29.

I²S DAC Crosstalk (Top Trace = Left to Right, Bottom Trace = Right to Left)



FREQUENCY (Hz) Figure 26.

PCM DAC Frequency Response (Handsfree Output)

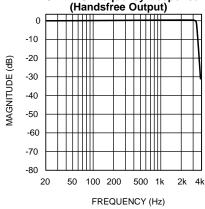


Figure 28.

PCM DAC Frequency Response (Headphone Output)

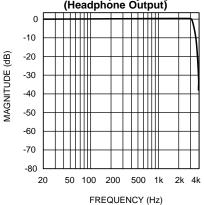


Figure 30.



Typical Performance Characteristics (1) (continued)

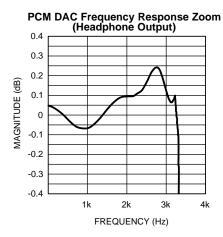
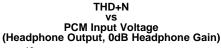
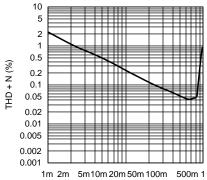


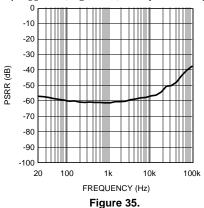
Figure 31.



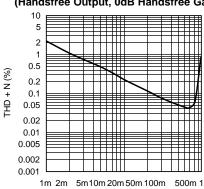


PCM INPUT VOLTAGE (FFS) **Figure 33.**

 $\begin{array}{c} \text{PSRR} \\ \text{vs} \\ \text{Frequency} \\ \text{(AV}_{DD} = 3V, R_L = 16\Omega, Headphone Output)} \end{array}$



THD+N
vs
PCM Input Voltage
(Handsfree Output, 0dB Handsfree Gain)



PCM INPUT VOLTAGE (FFS)

Figure 32.

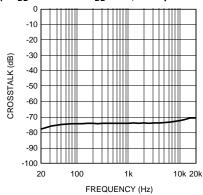


Figure 34.

 $\begin{array}{c} \text{PSRR} \\ \text{vs} \\ \text{Frequency} \\ (\text{AV}_{DD} = 3\text{V}, \, \text{R}_{L} = 32\Omega, \, \text{Headphone Output)} \end{array}$

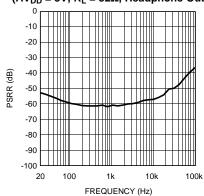


Figure 36.



Typical Performance Characteristics (1) (continued) PSRR PSR

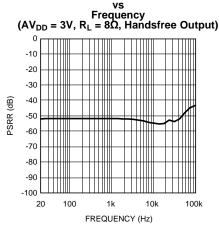
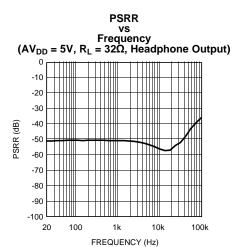


Figure 37.



 $THD+N \\ vs \\ Frequency \\ (AV_{DD}=3V,\,R_L=8\Omega,\,P_O=150mW,\,Handsfree\,\,Output)$

Figure 39.

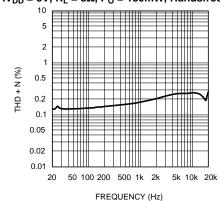


Figure 41.

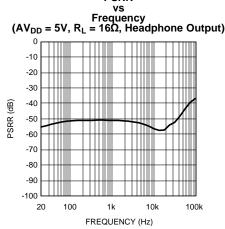
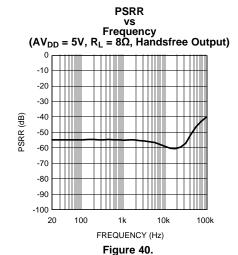


Figure 38.



THD+N vs Frequency (AV $_{DD}$ = 5V and AV $_{DD}$ = 3V, R $_{L}$ = 16 Ω , P $_{O}$ = 15mW, Headphone Output)

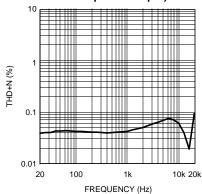
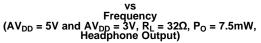


Figure 42.



Typical Performance Characteristics (1) (continued) THD+N



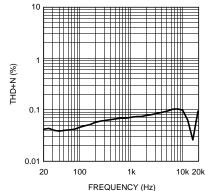


Figure 43.

 $THD+N \\ vs \\ Output \ Power \\ AV_{DD} = 3V, \ R_L = 16\Omega, \ f = 1kHz, \ Headphone \ Output)$

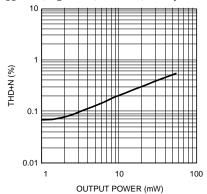


Figure 45.

 $THD+N \\ vs \\ Output \ Power \\ (AV_{DD}=5V \ and \ AV_{DD}=3V, \ R_L=32\Omega, \ f=1kHz, \ Headphone \\ Output)$

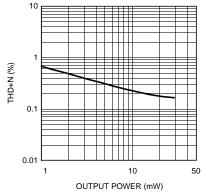


Figure 47.



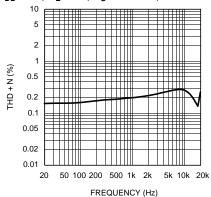
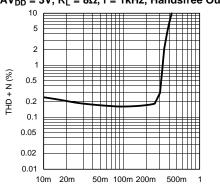


Figure 44.

$THD+N \\ vs \\ Output \ Power \\ (AV_{DD}=3V,\ R_L=8\Omega,\ f=1kHz,\ Handsfree\ Output)$



OUTPUT POWER (W) Figure 46.

 $THD+N \\ vs \\ Output Power \\ (AV_{DD}=5V \ and \ AV_{DD}=3V, R_L=16\Omega, f=1kHz, \ Headphone \\ Output)$

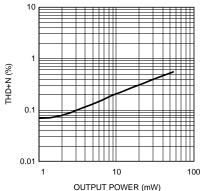
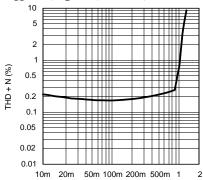


Figure 48.



Typical Performance Characteristics $^{(1)}$ (continued) $^{\text{THD+N}}$

 $\begin{array}{c} \text{VS} \\ \text{VS} \\ \text{Output Power} \\ \text{(AV}_{DD} = 5\text{V}, \, \text{R}_L = 8\Omega, \, \text{f} = 1\text{kHz}, \, \text{Handsfree Output)} \end{array}$



OUTPUT POWER (W)

Figure 49.



APPLICATION INFORMATION

REFERENCE DESIGN BOARD AND LAYOUT

LM4930ITL Board Layout

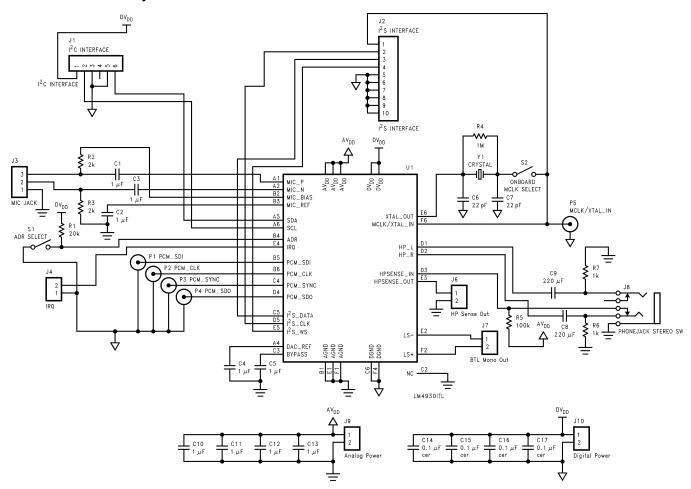


Figure 50. LM4930ITL Demo Board Schematic



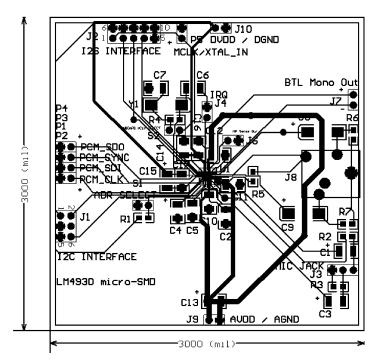


Figure 51. LM4930ITL Demo Board Composite View

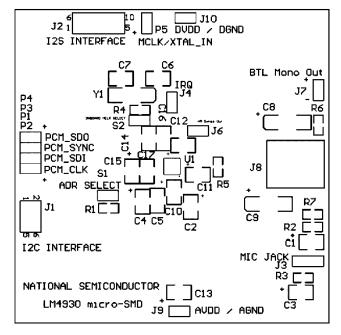


Figure 52. LM4930ITL Demo Board Silkscreen



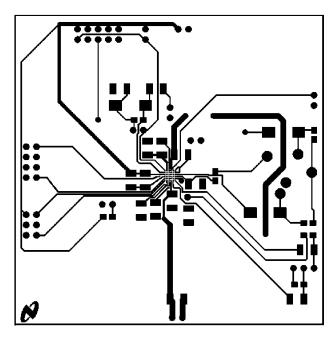


Figure 53. LM4930ITL Demo Board Top Layer

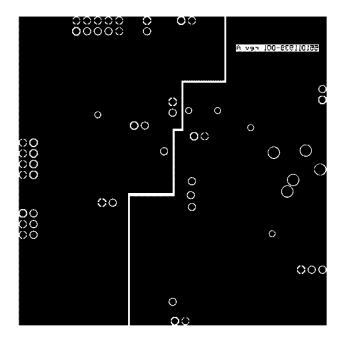


Figure 54. LM4930ITL Demo Board Bottom Layer



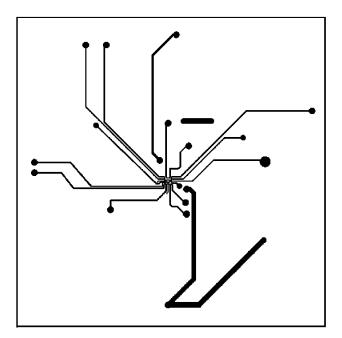


Figure 55. LM4930ITL Demo Board Inner Layer 1

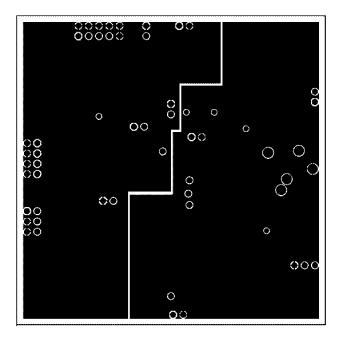


Figure 56. LM4930ITL Demo Board Inner Layer 2



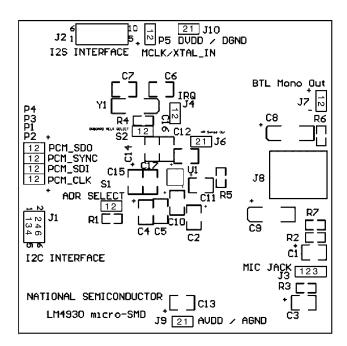


Figure 57. Pin Markings for LM4930ITL demo board

BILL OF MATERIALS FOR LM4930

Table 4. LM4930 Demo Board Bill Of Materials

Comment	Footprint	Designators
1k	0805	R6, R7
2k	0805	R2, R3
20k	0805	R1
100k	0805	R5
1M	0805	R4
22pF	1210	C6, C7
0.01µF cer	1210	C16, C17
0.1µF cer	1210	C14, C15
1µF	1210	C1, C2, C3, C4, C5, C10, C11, C12, C13
220µF	7243	C8, C9
CRYSTAL	7243	Y1
PHONE JACK STEREO SW STEREO HEADP	HONE JACK (3.5MM) J8	

Table 5. Two-Wire Control Interface (J1)

Pin	Function
1	DVDD
2	SCL
3	DGND
4	NC
5	DGND
6	SDA



Table 6. PCM Interface (P4, P3, P1, P2)

Header	Function
P1	PCM_SDI
P2	PCM_CLK
P3	PCM_SYNC
P4	PCM_SDO

Table 7. I2S Interface (J2)

Pin	Function
1	MCLK
2	I2S-CLK
3	I2S-DATA
4	12S-WS
5	DGND
6	DGND
7	DGND
8	DGND
9	DGND
10	DGND

Table 8. MIC Jack

Pin	Function
1	AGND
2	MIC-
3	MIC+

Table 9. Misc Jumpers and Headers DVDD/DGND (J10)

Pin	Function
1	DGND
2	AVDD

Table 10. Misc Jumpers and Headers AVDD/AGND (J9)

Pin	Function
1	AGND
2	AVDD

Table 11. Misc Jumpers and Headers MCLK/XTAL_IN (P5)

Pin	Function
1	DGND
2	MCLK/XTAL_IN



ADR SELECT (S1)

Jumper IN = LOW

Control interface responds to addresses 001000b (BASICCONFIG), 0010001b (VOICETESTCONFIG)), and 0010010b (GAINCONFIG)

Jumper OUT = HIGH

Control interface responds to addresses 111000b (BASICCONFIG), 1110001b (VOICETESTCONFIG)), and 1110010b (GAINCONFIG)

Table 12. HP Sense Out (J6)

Pin	Function
1	AGND
2	HPSense_Out

Table 13. IRQ (J4)

Pin	Function
1	DGND
2	IRQ

Onboard MCLK Select (S2)

Jumper IN = Onboard MCLK

Jumper OUT = External MCLK

LM4930ITL DEMO BOARD OPERATION

The LM4930ITL demo board is a complete evaluation platform, designed to give easy access to the control pins of the part and comprise all the necessary external passive components. Besides the separate analog (J9) and digital (J10) supply connectors, the board features seven other major input and control blocks: a two wire interface bus (J1) for the control lines, a PCM interface bus (P1-P4) for voiceband digital audio, an I2S interface bus (J2) for full-range digital audio, an analog mic jack input (J3) for connection to an external microphone, a BTL mono output (J7) for connection to an external speaker, a stereo headphone output (J8), and an external MCLK input (P5) for use in place of the crystal on the demoboard.

Two-wire Interface Bus (J1)

This is the main control bus for the LM4930. It is a two-wire interface with an SDA line (data) and SCL line (clock). Each transmission from the baseband controller to the LM4930 is given MSB first and must follow the timing intervals given in the Electrical Characteristics section of the datasheet to create the start and stop conditions for a proper transmission. The start condition is detected if SCL is high on the falling edge of SDA. The stop condition is detected if SCL is high on the rising edge of SDA. Repeated start signals are handled correctly. Data is then transmitted as shown in Figure 4. After the start condition has been achieved the chip address is sent, followed by a set write bit, wait for ack (SDA will be pulled low by LM4930), data bits 15-8, wait for ACK (SDA will be pulled low by LM4930) and finally the stop condition is given.

This same sequence follows for any control bus transmission to the LM4930. The chip address is hardwire selected by the ADR Select pin which may be jumpered high or low with its application at S1 on the demo board. The chip address is then given as a combination of the identifying bits for the LM4930 plus the 2-bit address of the desired control register (00b = BasicConfig, 01b = VoicetestConfig, 10b = GainConfig). Acceptable addresses are shown here in Table 14.



Table 14. LM4930 Control Bus Addresses

Address E	Bits		Register Address								
ADR = 0	ADR = 0										
6	5	4	3	2	1	0					
0	0	1	0	0	0	0					
0	0	1	0	0	0	1					
0	0	1	0	0	1	0					
ADR = 1											
1	1	1	0	0	0	0					
1	1	1	0	0	0	1					
1	1	1	0	0	1	0					

Data is sampled only if the address is in range and the R/W bit is clear. Data for each register is given in the System Control Registers section of the datasheet. Texas Instruments also features a special control board for quick evaluation of the LM4930 demo board with your PC. This is a serial control interface board, complete with header compatible with the interface header (J1) on the LM4930 board. This also features demonstration software to allow for complete control and evaluation of the various modes and functions of the LM4930 through the bus.

Pullup resistors are required to achieve reliable operation. 750Ω pullup resistors on the SDA and SCL lines achieves best results when used with TI's parallel-to-serial interface board. Lower value pullup resistors will decrease the rise and fall times on the bus which will in turn decrease susceptibility to bus noise that may cause a false trigger. The cost comes at extra current use. Control bus reliability will thus depend largely on bus noise and may vary from design to design. Low noise is critical for reliable operation.

PCM Bus Interface (P1, P2, P3, P4)

PCM_SDO (P4), PCM_SYNC (P3), PCM_SDI (P1), and PCM_CLK (P2) form the PCM interface bus for simple communication with most baseband ICs with voiceband communications and follow the PCM-1900 communications standard. The PCM interface features frame lengths of 16, 32, or 64 bits, A-law and u-law companding, linear mode, short or long frame sync, an energy-saving power down mode, and master only operation.

The PCM bus does not support a slave mode. It operates as a master only. Thus PCM_SYNC and PCM_CLK are solely generated by the LM4930. PCM_SYNC is the word sync line for the bus. It operates at a fixed frequency of 8kHz and may be set in the BASICCONFIG register (bit 5 PCM_LONG) for short or long frame sync. A short frame sync is 1 PCM CLK cycle (PCM LONG=0), a long frame sync is 2 PCM CLK cycles long (PCM LONG=1). A long sync pulse is also delayed one clock cycle relative to a short sync pulse. This is illustrated in Figure 5. PCM CLK is the bit clock for the bus. It's frequency depends on the number of 16-bit frames per sync pulse and can be 128kHz, 256kHz, 512kHz.

The other two lines, PCM_SDO and PCM_SDI, are for serial data out and serial data in, respectively. The type of data may also be set in the BASICCONFIG register by bits 6 and 7. Bit 6 controls whether the data is linear or companded. If set to 1, the 8 MSBs are presumed to be companded data and the 8 LSBs are ignored. If cleared to 0, the data is treated as 2's complement PCM data. Bit 7 controls which PCM law is used if Bit 6 is set for companded (G711) data. If set to 1, the companded data is assumed to be A-law. If cleared to 0, the companded data is treated as µ-law.

Bits 8:9 of the BASICCONFIG register set the PCM_SYNC_MODE settings. This controls the number of 16 bit frames per sync pulse. The feature allows the LM4930 to function harmoniously with other devices or channels on the PCM bus by adjusting the number of 16 bit frames per sync pulse to 1 (00b), 2 (01b), or 4 (10b). The LM4930 will transmit PCM data in the first frame and then tri-state the PCM_SDO pin on later frames.

In addition, the LM4930 provides control to allow the PCM_CLK and PCM_SYNC clocks to continue functioning even when the LM4930 is in Standby mode. By setting bit 10 of the BASICCONFIG register to 1 PCM_ALWAYS_ON is enabled and the LM4930 will continue to drive the PCM clock and sync lines when in Standby mode. This bit should be set if another codec is using the PCM bus. Powerdown mode will disable these outputs.

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I2S Interface Bus (J2)

The I2S standard provides a uni-directional serial interface designed specifically for digital audio. For the LM4930, the interface provides access to a 48kHz, 16 bit full-range stereo audio DAC. This interface uses a three port system of clock (I2S_CLK), data (I2S_DATA), and word (I2S_WS). The clock and word lines can be either master or slave as set by bit 11 in the BASICCONFIG register.

A bit clock (I2S_CLK) at 32 or 64 times the sample frequency is established by the I2S system master and a word select (I2S_WS) line is driven at a frequency equal to the sampling rate of the audio data, in this case 48kHz. The word line is registered to change on the negative edge of the bit clock. The serial data (I2S_DATA) is sent MSB first, again registered on the negative edge of the bit clock, delayed by 1 bit clock cycle relative to the changing of the word line (typical I2S format - see Figure 6).

The resolution of the I2S interface may be set by modifying the I2S_RES bit (bit 12) in the BASICCONFIG register. If set to 1, the LM4930 operates at 32 bits per frame (3.072MHz). If cleared to 0, then 16 bits per frame is selected (1.536MHz). This has a corresponding effect on the bit clock.

The I2S Interface Bus also provides for an additional MCLK connection to an external device from the LM4930 demo board. This may be used in conjunction with Texas Instruments' SPDIF->I2S Conversion Board for quick evaluation. This board features a connection header that interfaces with pins 1-5 of the I2S Interface Bus. Pins 6-10 are provided as digital ground references for the case of discrete connections.

MCLK/XTAL IN (P5)

This is the input for an external Master Clock. The jumper at S2 must be removed (disconnecting the onboard crystal from the circuit) when using an external Master Clock.

BTL Mono Out (J7)

This is the mono speaker output, designed for use with an 8 ohm speaker. The outputs are driven in bridge-tied-load (BTL) mode, so both sides have signal. Outputs are normally biased at one half AVDD when the LM4930 is in active mode.

Additionally, if the CLASS bit is set to 1 in the VOICETESTCONFIG register (bit 0) the BTL mono output is internally configured as a buffer amplifier designed for use with an external class D amp.

Stereo Headphone Out (J8)

This is the stereo headphone output. Each channel is single-ended, with 220uF DC blocking capacitors mounted on the demo board. The jack features a typical stereo headphone pinout.

A headphone sense pin is provided at J6. This pin provides a clean logic high or low output to indicate the presence of headphones in the headphone jack. A common application circuit for this is given in the Reference Board Schematic shown in Figure 50. In this application HPSENSE_IN is pulled low by the 1k ohm resistor when no headphone is present. This gives a corresponding logic low output on the HPSENSE_OUT pin. When a headphone is placed in the jack the 1k ohm pull-down is disconnected and a 100k ohm pull-up resistor creates a high voltage condition on HPSENSE_IN. This in turn creates a logic high on HPSENSE_OUT. This output may be used to reliably drive an external microcontroller with headphone status.

MIC Jack (J3)

This jack is for connection to an external microphone like the kind typically found in mobile phones. Pin 1 is GND, pin 2 is the negative input pin, and pin 3 is the positive pin, with phantom voltage supplied by MIC_BIAS on the LM4930.

IRQ (J4)

This pin provides simple status updates from the LM4930 to an external microcontroller if desired. IRQ is logic high when the LM4930 is in a stable state and changes to low when changing modes. This can also be useful for simple software/driver development to monitor mode changes, or as a simple debugging tool.



BASIC OPERATION

The LM4930 is a highly integrated audio subsystem with many different operating modes available. These modes may be controlled in the BASICCONFIG register in bits 3:0. These mode settings are shown in the BASICCONFIG register table and are described here below:

Powerdown Mode (0000b)

Part is powered down, analog outputs are not biased. This is a minimum current mode. All part features are shut down.

Standby Mode (0001b)

The LM4930 is powered down, but outputs are still biased at one half AVDD. This comes at some current cost, but provides a much faster turn-on time with zero "click and pop" transients on the headphone out. Standby mode can be toggled into and out of rapidly and is ideal for saving power whenever continuous audio is not a requirement. All other part functions are suspended unless PCM_ALWAYS_ON (bit 10 in BASICCONFIG register) is enabled, in which case PCM_CLK and PCM_SYNC will continue to function.

Mono Speaker Mode (0010b)

Part is active. All analog outputs are biased. Audio from the voiceband codec is routed to the mono speaker out. Stereo headphone out is silent.

Headphone Call Mode (0011b)

Part is active. All analog outputs are biased. Audio from voiceband codec is routed to the stereo headphones. Both left and right channels are the same. Mono speaker out is silent.

Conference Call Mode (0100b)

Part is active. All analog outputs are biased. Audio from the voiceband codec is routed to the mono speaker out and to the stereo headphones.

L+R Mixed to Mono Speaker (0101b)

Part is active. All analog outputs are biased. Full-range audio from the 16bit/48kHz audio DAC is mixed together and routed to the mono speaker out. Stereo headphones are silent.

Headphone Stereo Audio (0110b)

Part is active. All analog outputs are biased. Full-range audio from the 16bit/48kHz audio DAC is sent to the stereo headphone jack. Each channel is heard discretely. The mono speaker is silent.

L+R Mixed to Mono Speaker + Stereo Headphone Audio (0111b)

Part is active. All analog outputs are biased. Full-range audio from the 16bit/48kHz audio DAC is sent discretely to the stereo headphone jack and also mixed together and sent to the mono speaker out.

Mixed Mode (1000b)

Part is active. All analog outputs are biased. This provides one channel (the left channel) of full range audio to the mono speaker out. Audio from the voiceband codec is then sent to the stereo headphones, the same on each channel.

Mixed Mode (1001b)

Part is active. All analog outputs are biased. Mixed voiceband and full-range audio (left channel only) is sent to the mono speaker out. Audio from the voiceband codec only is sent to the stereo headphones, the same on each channel.



Mixed Mode (1010b)

Part is active. All analog outputs are biased. Audio from the voiceband codec is sent to the mono speaker out. The left channel only of the full range audio is then sent to both the left and right channels of the stereo headphone out.

REGISTERS

The LM4930 starts on power-up with all registers cleared in Powerdown mode. Powerdown mode is the recommended time to make setup changes to the digital interfaces (PCM bus, I2S bus). Although the configuration registers can be changed in any mode, changes made during Standby or Powerdown prevent unwanted audio artifacts that may occur during rapid mode changes with the outputs active. The LM4930 also features a soft reset. This reset is enabled by setting bit 4 of the BASICCONFIG register.

The VOICETESTCONFIG register is used to set various configuration parameters on the voiceband and full-range audio codecs. SIDETONE_ATTEN (bits 4:1) refers to the level of signal from the MIC input that is fed back into the analog audio output path (commonly used in headphone applications and killed in hands-free applications). Setting the AUTOSIDE bit (bit 5) automatically mutes the sidetone in voice over mono speaker modes so feedback isn't an issue.

Quick mute functions are also located in this register, with bits 13:15 muting the mono speaker amp, the headphone amp, and the mic preamp respectively.

This register also has a CLOCK_DIV bit (bit 6) which, if set, allows for the use of a 24.576MHz clock instead of the default 12.288MHz.

The GAINCONFIG register is used to control the gain of the mono speaker amp, the headphone amp, and the mic preamp. This allows flexible mono speaker gains from -34.5dB to +12dB in 1.5dB steps, headphone amp gains of -46.5dB to 0dB in 1.5dB steps, and mic preamp gains of 17dB to 47dB in 2dB steps. Gain levels may be modified in any mode, but may wait for a zero cross detect in the DAC to eliminate volume control artifacts. This wait for zero cross may be disabled by setting the ZXD_DISABLE bit (bit 7) in the VOICETESTCONFIG register to allow immediate changes.

ANALOG INPUTS AND OUTPUTS

The LM4930 features an analog mono BTL output for connection to an 8Ω external speaker. This output can provide up to 1W of power into an 8Ω load with a 5V analog supply. A single-ended stereo headphone output is also featured, providing up to 30mW of power per channel into 32Ω with a 5V analog supply.

A Headphone Sense output is provided on J6 for connection to an external controller. This pin goes high when a heaphone is present (when used as shown in Figure 50) and will function in all modes independent of other operations the LM4930 may be currently processing.

The MIC Jack input (J3) provides for a low level analog input. Pin 3 provides the power to the MIC and the positive input of the LM4930. Gain for the MIC preamp is set in the GAINCONFIG register.

SNAS212C - JULY 2003 - REVISED MAY 2013



REVISION HISTORY

Cł	nanges from Revision B (May 2013) to Revision C	Page
•	Changed layout of National Data Sheet to TI format	33



PACKAGE OPTION ADDENDUM

10-Dec-2020

PACKAGING INFORMATION

www.ti.com

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
LM4930ITLX/NOPB	ACTIVE	DSBGA	YZR	36	1000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-30 to 85	G B6	Samples
LM4930LQ/NOPB	ACTIVE	WQFN	NJN	44	250	RoHS & Green	SN	Level-2-260C-1 YEAR	-30 to 85	L4930LQ	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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10-Dec-2020

PACKAGE MATERIALS INFORMATION

www.ti.com 20-Sep-2016

TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

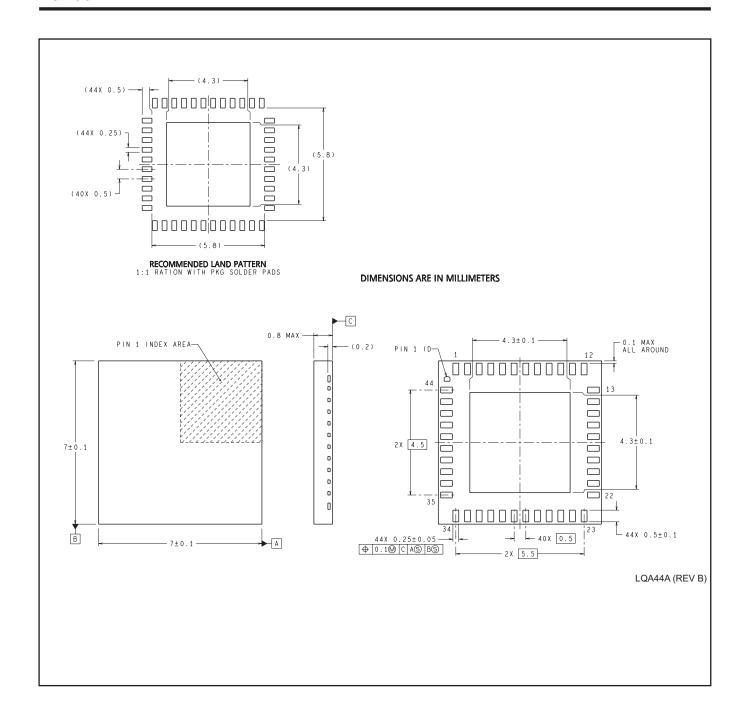
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM4930ITLX/NOPB	DSBGA	YZR	36	1000	178.0	12.4	3.43	3.59	0.76	8.0	12.0	Q1
LM4930LQ/NOPB	WQFN	NJN	44	250	178.0	16.4	7.3	7.3	1.3	12.0	16.0	Q1

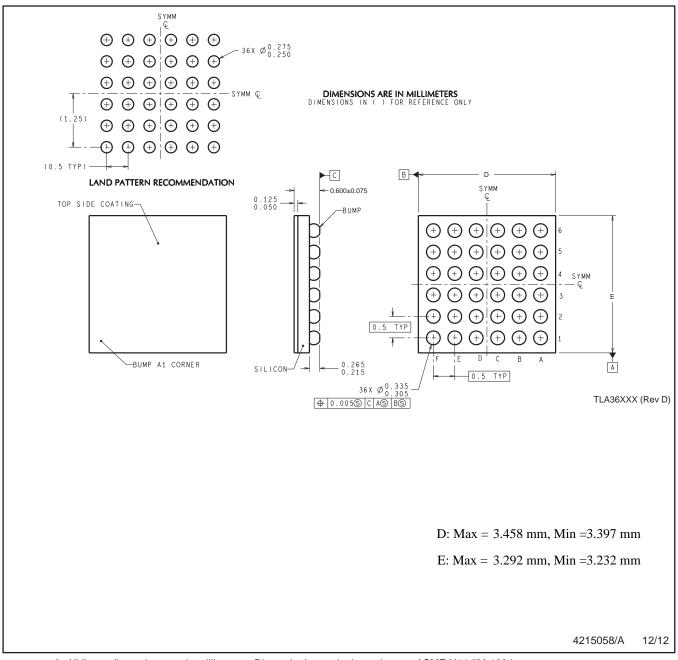
www.ti.com 20-Sep-2016



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
LM4930ITLX/NOPB	DSBGA	YZR	36	1000	210.0	185.0	35.0	
LM4930LQ/NOPB	WQFN	NJN	44	250	210.0	185.0	35.0	





NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994. B. This drawing is subject to change without notice.



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